- 1/61 -

JUICE Radio and Plasma Wave Instrument

Low Frequency Receiver digital interface

JUI-IAP-RPWI-LF-DIF Issue: 01, Rev. 154

 $\overline{}$

- 2/61 -

Change Record

 $\overline{}$

- 3/61 -

Table of Contents

 $\overline{}$

 $\overline{}$

1 Applicable and reference documents

Applicable documents

This document responds to the requirements of the documents listed in the following table:

Reference documents

The present document refers to the documents listed in the following table:

2 LFR data products (FPGA)

LFR FPGA shall produce a combination of the following data products as configured by software in the Data product configuration register (0x0006). Any combination of the data product is allowed with the exception that CWF and WFS cannot be enabled simultaneously.

Waveform snapshots (WFS): Waveform snapshots are blocks of 8 x 128 x N. CWF. frames (see register WFS length 0x9) digitized at 48.828 ksps or 24.414 ksps buffered internally in the LF and sent after the collection has ended.

Continuous waveform (CWF): A continuous stream of samples digitized at 48.828 ksps or 24.414 ksps (8 channels).

Decimated waveform (DWF): A continuous stream of decimated data digitized at 762 sps.

Spectral matrices (SM): Blocks of 8 x 8 x (number of frequency bins) 64-bit numbers sent at a low cadence (at most once per second).

Raw FFT (FFT): A diagnostic product allowing to transmit complex FFTs for al channels continuously. For test only, not to be used in flight.

FFT Sum (FSUM): Component sum power spectra calculated at full time resolution (every 1024 samples). One spectrum of 1024 frequencies transmitted per packet. Two spectra can be configured obtained by summing components specified in register 0x18 (register name: FFT bins summation mask 1 and mask2). The output 16 bits samples are custom float numbers composed of 6 bits exponent of base2 (MSbits) and 10 bits of mantissa (LSbits). This number is sum of selected components. Each component have their real part squared summed with imaginary part squared (RE^2+IM^2). The sample decimal value is decoded from sample bits as:

- exponent = sample bits $(15$ downto 10)
- mantissa = sample bits $(9$ downto 0)
- sample value = mantissa $<<$ exponent

DCFG: A special diagnostic packet containing current full configuration and status of LF (addresses 0x0000 to 0x01FF) of the LF configuration memory, with actual time, except 0x12 register – Parity error counter register. The space between the end of LF configuration map (0x01FF) and the end of the packet is filled by a testpattern (0x0, 0x555555555, 0xAAAAAAAA, 0xFFFFFFFFF) varying every 32 bits. The rest of the packet Transmitted as a single frame on request (when register 0x0006 is written with bit 15 set to 1).

3 Address map

All commanding of LF and readout of status information by the DPU is performed by reading and writing the address space exported by LFR. Map of the LF address space visible to DPU over link is detailed. Only the first 0x0200 words are used.

JUICE RPWI LF digital interface

 $\overline{}$

Ref: JUI-IAP-RPWI-LF-DIF Issue: 01 Revision: 154 Date: **04/10/2023**

- 8/61 -

4 Description of individual registers

The PPS counter is a 4-bit counter incremented by LF with every PPS pulse. The skew is incremented internally with the clock frequency of 781.25 kHz (50 MHz / 64). The skew is reset to 0 with every PPS pulse, when PPS counter is incremented.

It is possible to read the PPS counter value from register 0x0000, as long as it is least 1us after PPS signal. To read PPS it is not necessary to write bit 15.

To read the skew time, a two-step procedure is needed:

- Write the register 0x0000 with a value of 0x8000.
- Read the content of registers 0x0000 and 0x0001.

Register 0x0000 can also be used to set the PPS value. PPS value is set by writing this register, with bit 14 set to 1 and bit 15 set to 0. The value at bits 2-5 is then set as the new PPS value. Other bits are ignored. **The new value is applied when LF receives the next PPS pulse, not immediately.**

The following 4 registers are used to configure hardware switches on LFR. Each bit corresponds to one switch.

 $\overline{}$

 $\overline{}$

- 11/61 -

Note: if the data products bits are changed before the CHCFGREQ bit is cleared then the invalid command flag (register 0x10) is set

Note: LFR accommodates only one buffer for WFS data. If the buffer is filled up then it must be sent out before a new data acquisition begins – thus the configured snapshot period can be prolonged.

JUICE RPWI LF digital interface

 $\overline{}$

 \mathbf{I}

- 13/61 -

Note: useful for SRAM read/write test. RAM address starts at 0xa0000000 and occupies 8MB (up to 0xa07fffff).

FFT bins summation mask 1 and mask2 (address 0x0018): FSUM product configuration

- 15/61 -

5 Commanding of LFR

When powered on LFR will start up in standby mode, where:

- Analog sections are powered down
- No packets are transmitted

To configure LFR into science mode, the software has to perform the following steps:

- 1) Configure hardware switches (write words at address 0x0002, 0x0003,0x0004, 0x0005, 0x000B)
- 2) If SMX product is desired (corresponding bit will be set in register 0x0006):
	- Set number of spectral bins (NFB) in register 0x000F.
	- Upload and configure NFB spectral bins (0x0100 to 0x01FF).
	- Configure mask of bins to include ($0x0020$ to $0x005F$) if desired. Otherwise a default is used (all bins will be included in averaging).
	- Configure number of averaged spectra in register 0x000A
- 3) Write the PPS value to start data acquisition in register 0x0007
- 4) If WFS snapshot acquisition is required, set the WFS period in register 0x0008.
- 5) Write register 0x0006, setting the bits corresponding to the requested data products. This actually enables the science mode. The acquisition starts on the next PPS signal.

Changing configuration / re-synchronizing: Every time register 0x0006 is written (with bit CHCFGREO = 1). New configuration is applied and data acquisition is stopped and re-started on the PPS pulse specified in register 0x0007. This has to be done for every change in instrument configuration. If no configuration changes are made before writing to 0x0006, the same configuration applies, but data acquisition is resynchronized to the specified PPS pulse.

Generating SCM calibration signal: LF can transmit a synthetic signal to SCM with the following configuration:

 Upload the waveform to be transmitted as 1024 samples (unsigned 16-bit integers) by writing to addresses 0x200-0x5FF. This waveform will later be played back through the DAC (DAC range is

- 16/61 -

12 bits, so full useful range is between ~100 and 4000. The full range is translated to an output voltage range of about 0 to +5V).

- Configure all LF registers in the desired science configuration same as above.
- Before enabling the configuration by register 0x0006, write register 0x0C to set bit 2 to 1.
- Enable science mode as normal. When the data acquisition starts on the next PPS pulse, the SCM signal is generated immediately.
- The Uploaded waveform is played back in the following manner:
	- o It is played to the DAC at a frequency of 524 kHz (2^19 Hz)
	- **O** Initially every $64th$ sample is sent to output and this is repeated 4 times.
	- o Next every 32th sample is sent to output and this is repeated 4 times.
	- o Next every 16th sample is sent to output and this is repeated 4 times.
	- o ….
	- o Full waveform is played at DAC frequency of 524 kHz (four times).
	- o Next the full waveform is played (4 times) at half of the initial sampling frequency (at 262 kHz)
	- o Then the sampling frequency is halved at every step until the sampling frequency gets to 256 Hz.
- This way, if one uploads to LF one period of a sine wave (centered around 2048 integer value), we get a logarithmic frequency sweep starting at 32768 Hz and ending at 0.25 Hz. At each step, four wave periods are transmitted and frequency is halved at every step.

 $\overline{}$

- 17/61 -

6 LFR registers to be used for instrument HK packet

7 Data transfer and format

All data is transmitted as data frames of 1024 16-bit words + header of 10 16-bit words using the push transfer type of the serial link. Each data frame has an identical format described below (section Data frame layout). The data product contained in this frame is determined by the content of the first 16-bit word of the header.

- Waveform snapshots are always sent as frames of 8x128 16-bit words. Unused channels contain random data. The Acquisition time correspond to the first sample of snapshot and only the PPS value is valid (skew should be 0)
- DWF and CWF sent as frames of 8x128 16-bit words. The Acquisition time corresponds to the first sample in the frame.
- Spectral matrices are sent as 8x8 (64) sets of 64-bit numbers corresponding to a single frequency (512 bytes per frequency). NFB frequency bins are transmitted sequentially, 2 frequencies per frame. Note: for SM, the Acquisition time correspond to the first **sample** of the last FFT frame in the matrix.

Special **DCFG** diagnostic packet (data id == 0x8000) is sent by LF on request, when 1 is written to DCFG bit in register 0x0006. This packet contains the dump of the configuration memory of LF (addresses 0x0000 to 0x01FF). This packet has a standard dateframe layout with a standard header and length.

8 Introduction

This document provides a specification of communication protocol for the Low Frequency receiver board of the RPWI instrument for JUICE. The document describes both the interface between the FPGA and DPU as well as the configuration and structure of the data products transmitted by RPWI to the spacecraft.

9 Data frame layout

This table specified the format of the data frame. All data products are transmitted in frames formatted according to this specification using the push operation. The length of the data frame shall probably be fixed to $1024 + 12 = 1036$ 16-bit words (2072 bytes).

 $0x000C$ Start of data R \overline{R} N x 16 bit

10 Description of individual fields

The acquisition time (time when data was measured) is encoded in 2 16-bit words at 0x1002 and 0x1003 (see the format below). Usually, for DWF, CWF and WFS waveform products, the time corresponds to the first sample in the given data frame. A special case are Spectral Matrices, where the time corresponds to the first sample of the last FFT accumulated in the matrix. So the time corresponding to the beginning of the averaging has to be reconstructed by subtracting NAS*1024*cwf_sampling_period from the acquisition time (where NAS is the content of register 0x000A, number of averaged FFT spectra – 1, and the cwf_sampling_period is either 1/48828.125 or 1/24414.063 sec, depending on decimation setting).

 $\overline{}$

- 20/61 -

11 Datarates

This section describes the data products generated by the LF receiver toward the DPU (not the actual RPWI products, which are further processed in software). The LF receiver will produce the telemetry at the following rates:

CWF (Continuous waveform): One frame of 8x128 samples + header transmitted every 2.6 milliseconds for 48.8 ksps sampling rate (~382 packets per second) continuously. For the reduced sampling rate of 24.4 ksps, the datarate is decrease by half to one frame per 24.4 ksps. When CWF is produced, waveform snapshots are not collected.

WFS (Waveform snapshots): This product is sent in the same manner as the CWF product, with the difference that it is not transmitted continuously. LF will transmit a series of frames corresponding to one snapshots (at most 6 MB of data = 8x393216 samples) followed by a gap when no data is transmitted. The rate of packet transmission can be decreased if necessary.

DWF: One frame of 8x128 samples + header transmitted every 168 milliseconds (6 packets per second). If enabled, this product is transmitted continuously.

SM (spectral matrices):

- The rate of spectral matrices depends on the configuration parameters NFB (number of frequency bins) and NAS (number of averages spectra -1).
- Spectral matrices are sent as 8x8 (64) sets of 64-bit numbers corresponding to a single frequency (512 bytes per frequency). NFB frequency bins are transmitted sequentially, 4 frequencies per frame. The size of each frame is 2048 bytes (or 1024 16-bit words) + header, same as for the waveform products.
- One full matrix (512*NFB bytes = NFB/4 packets) is sent every 20.9*(NAS+1) milliseconds.
- Minimum value for NAS is 2, corresponding to a rate SM_RATE of one matrix every 63 milliseconds (~16 matrices per second).
- A limit is imposed on the product SM_RATE * NFB < 512. This implies that for NAS=2 (SM_RATE $=$ \sim 16 matrices per second), the maximum number of bins is 32 and for NFB = 128 at most 4 matrices can be produced every second. This worst case rate corresponds to 128 SM packets every second.
- Typical SM rate will be one matrix of 32 frequencies every 4 seconds (2 packets per second).

FFT Sum (FSUM): When enabled the component sum power spectra is calculated every 1024 CWF samples that takes 21ms and 2072 bytes of the LFR packet. If two FFT sum power spectra are configured then two LFR packets are generated every 21 ms. For the datarate one FFT sum power spectra consumes ~98,7 kBps (48.8 ksps), for two FFT sum power spectra it consumes ~197,3 kBps (97.5 ksps).

 $-23/61$

12 Memory organization (FPGA)

The LFR SRAM memory of 8 Mbytes (8388608 bytes) is organized as follows:

Table 12-1 LFR SRAM organization (values are in HEX format). Note that the address used by the SRAM controller is prefixed with 0xA0 (Address 0 = 0xA0000000)

13 FPGA side processing of spectral matrices

Notation used:

- NFB is number of frequency bins between 4 and 128. Uploaded via TC.
- \bullet n = index of output frequency bin (n=1...NFB).
- \bullet i = row index of spectral matrix (i=1...8)
- $j =$ column index of spectral matrix ($j=1...8$)
- $f = FFT$ frequency index. Index to the raw FFT ($f = 1...1024$)
- \bullet $F_c(f)$ = FFT of channel "c", frequency bin f. Complex number 16-bits. Stored in SRAM.
- bs(n) first frequency in output bin number n. bs(n) is an index to $F_c(f)$ and is between 1..1024. bs(n) is uploaded as a table via TC at LFR initialization.
- be(n) last frequency in output bin number n. be(n) is an index to $F_c(f)$ and is between 1..1024. be(n) is uploaded as a table via TC at LFR initialization.
- ASM_{ii}(n) = output accumulated spectral matrix. For each output bin "n" it is $8 \times 8 \times 64$ bits. Stored in external RAM. Real numbers. Real part of cross spectrum for j > i, imaginary for i \le j and auto spectrum (real) for $i = j$.
- include mask = bitmask of 1024 bits. If a bit is set to 1, the corresponding FFT bin is included in averaging in frequency.

Algorithm for SM calculation:

```
for n = 1 to NFB
    if (first_accumulated_fft)
        ASM_{ii}(n) = 0 for i=1..8, j=1..8else
        Load ASM_{11}(n) from external RAM for i=1..8, j=1..8
    end
```

```
for f = bs(n) to be(n)
         if (include\_mask(f) == 0)continue
         end
         load F_c(f) for all c=1..8 from RAM to internal buffer.
         for i=1..8
             for j=1..8
                 if (j \ge i)ASM_{ij}(n) += Re[F_i(f)]*Re[F_j(f)] + Im[F_i(f)]*Im[F_j(f)]else
                    ASM<sub>ij</sub>(n) += Im[F<sub>i</sub>(f)]*Re[F<sub>i</sub>(f)] - Re[F<sub>i</sub>(f)]*Im[F<sub>i</sub>(f)]
                 end
             end
         end
     end
     Write ASM_{ij}(n) back to external RAM for i=1..8, j=1..8end
```
14 LF software configuration

14.1 Processing compression encoding (PCE) configuration – LF main configuration structure

This structure contains a complete configuration of LF, excluding the spectral bin configuration tables described in next section.

 $\overline{}$

- 26/61 -

 \mathbb{I}

- 28/61 -

Total size = 120 bytes.

 $\overline{}$

Description of individual fields follows:

 $-29/61$

14.2 Spectral bin tables (SB_TABLE)

The spectral bin tables specify the edges of frequency bins for averaging of spectral products. These shall be stored in DPU MRAM (changeable by TC) and referred to by an index (SB_INDEX) from the main configuration structure. When LF board is configured, the table corresponding the specified index shall be uploaded in the LF board using a sequence of write commands.

Each table is 516 bytes long. The DPU shall be able to store up to 16 such tables, each with the following structure:

Total size: 516 bytes.

14.3 Include frequency mask table (EFM table)

The spectral bin tables shall be stored in DPU MRAM (changeable by TC) and referred to by an index (EFM_INDEX) from the main configuration structure. Each table is 132 bytes long. The DPU shall be able to store up to 8 such tables, each with the following structure:

Total size: 132 bytes.

15 LF TM packets (SW interface to OBC)

The following LF science TM data products can be generated.

15.1 TM_LF_RAW

This data product is transmitted in RAW mode and is basically a raw LF hardware frame with RPWI header. Designed for debug/calibration.

- 31/61 -

15.2 TM_LF_RSWF

 $\overline{}$

Periodically collected waveform snapshots sampled at 48.8 ksps or 24.4 ksps, divided into multiple packets. Aux header only in the first packet.

- 32/61 -

15.3 TM_LF_TSWF

Triggered waveform snapshot sampled at 48.8 ksps or 24.4 ksps, divided into multiple packets. Aux header only in the first packet.

TRIGGER_INFO 16bit word contains the following:

- 33/61 -

15.4 TM_LF_DWF

Continuous waveform sampled at 763 Hz or optionally decimated down to 763/2^N Hz.

15.5 TM_LF_SM

On-board calculated full spectral matrix (8 x 8 or reduced to smaller dimensions).

 $\overline{}$

Spectral matrix data block per frequency bin (BLOCK_SIZE bytes each)

15.6 TM_LF_BP0

 $\overline{}$

Simple spectral product for selective downlink. Only contains E and B power spectra calculated in software from spectral matrices. Not split in multiple packets.

15.7 TM_LF_BP1

 $\overline{}$

Extended spectral product, to be used as quicklook for selective downlink, but also for science. Not split in multiple packets.

 $\overline{}$

- 38/61 -

Data block per frequency bin (16 bytes per block)

15.8 TM_LF_DWFS

 $\overline{}$

A waveform snapshot created from the DWF product, divided into multiple packets. Aux header only in the first packet.

- 40/61 -

15.9 TM_LF_BP2

Another simple spectral product for selective downlink. This contains very reduced power spectra and wave polarization parameters. Not split in multiple packets.

Data block per frequency bin (4 bytes per block)

15.10 TM_LF_STAT

A very short TM packet containing triggering results, including dust and wave counters, peak/RMS amplitudes etc. Not split in multiple packets.

15.11 Common header data

HW_SWITCHES_ARTEFACTS: A compressed bitmask containing complete information about LF hardware switches and data artefacts (overflows). Size of this bitmask is 6 bytes (48 bits).

RW_FREQS: 8-byte array where each byte indicates the FFT bin masked due to the reaction wheel signature. There are 4 RWs and for each RW we can mask the fundamental and the 8-th harmonic, hence up to 8 frequencies can be basked out (8 bytes are necessary).

The value of the byte N_rw is a frequency bin corresponding to the original 2048 point FFT (masked freq = f_samp * N_rw / 2048), where f_samp is LFR sampling frequency (either 48828 Hz or 24414 Hz). There are two special values: 0xFF indicates no mask is used and 0xFE is an invalid value (generally indicating and error).

16 DPU-side software processing of LF data

This section describes the handling of the LF data in the DPU software and the LF software operation in the normal operational mode (LF_NOMINAL_SCI submode or the SCINECE mode). LF software also supports other modes (LF_RAW and LF_RFT_FFT), where the processing is effectively disabled, but these are only used for testing.

In this section and below, the variables in *bold-italic* refer to configuration parameters in the PCE configuration structure.

16.1 Basic operation and configuration of LF software

The operation of LF in SCIENCE mode (NOMINAL_SCI submode) is configured by the following data structures modifiable by telecommands.

Two structures which are a part of the LF configuration selected either by TC or by the sequencer

HD_Config structure: Structure directly configuring some LF FPGA registers (For the description of the registers, see Section [4\)](#page-7-0). In normal SCIENCE mode, only a subset of the register values in this structure is used:

- o HW_SWITCH_CONFIG1..4 sets the hardware switches of LF board (registers 0x02 to 0x05)
- o PWR_CTRL configures register 0x0B (controls power of SCM and LF board and other hardware features).
- o NAS Number of spectral matrices to average in FPGA (register 0x0A).
- o lf_submode should be set to LF_NOMINAL_SCI.

The rest of values is ignored and the registers are set with values calculated by the software from the settings in the PCE structure. For the description of the registers, see Section [4.](#page-7-0)

 PCE_Config structure: The main configuration structure, defining most of LF software behaviour and data generation. See Section [14.1](#page-24-1) for description.

Multiple parameter tables needed for data processing:

- Spectral bin tables (SB TABLE): These tables configure the output "reduced" frequency bins of the spectral products. Only one table is active at a time, selected by an index in the PCE_Config structure. See Section [14.2](#page-28-0) for description.
- Include frequency mask table (EFM_TABLE): These tables configure the which frequency bins should be included in the averaging of the spectral products. Only one table is active at a time, selected by **SM_INCLUDE_MASK_INDEX** in PCE_Config.
- There are more tables …. **TBW**

16.2 High resolution waveform snapshots (TM_LF_RSWF and TM_LF_TSWF)

The TM_LF_RSWF (Regular Science Waveform Snapshot) data product returns multi-component waveform snapshots taken periodically in regular intervals specified by *WFS_PERIOD*. The TM_LF_TSWF product provides the same snapshots, but instead of being captured periodically, they are selected based on one of the triggering algorithms described below. The TM_LF_RSWF and TM_LF_TSWF products are mutually exclusive, only one can be enabled at a time.

Software will need to select a subset of the components specified by a bitmask *WFS_CHANNEL_MASK* in the configuration structure. Only the selected components will be put in the TM packet. The snapshot length (number of snapshots per channel) is specified in *WFS_LENGTH* in multiples of 128 points.

16.3 Decimated Waveform at 763/2^N sps (TM_LF_DWF and TM_LF_DWFS)

LF supports two decimated data products generated from the slow data sampled by the FPGA at 763 sps.

 TM_LF_DWF is a continuous time series than can be decimated from the original 763 Hz by a factor of 2^N configured by *DWF_DECIMATION* using an FIR filter. This is used to generate the 24 Hz continuous data. Any set of field components can be selected by *DWF_CHANNEL_MASK*. This product is always transmitted in individual data packets, each containing 128 samples**.**

 TM_LF_DSWF is a multi-component time series sampled always at 763 Hz and *DWFS_CHANNEL_MASK* configures the E/B components to be included. This product is generated in the form of "snapshots" similar in format to TM_LF_RSWF. The length of the DSWF snapshots in samples per component is set by the *DWFS_LENGTH* parameter (in multiples of 128). DSWF product. TM_LF_DWFS can be generated either periodically (similar to TM_LF_RSWF) or in response to a trigger.

To enable periodic DWFS generation, the bit CFG_EXTRAS_DWFS_NO_SEQ has to be set in the EXTRA_SETTINGS variable and the period (start to start) is set via DWFS_PERIOD in multiples of 128 samples per channel. This can be used to generate a continuous DSWF waveform (such as in the insitu burst mode) by setting *DWFS_PERIOD* equal to *DWFS_LENGTH (e.g. to 0x20).*

For description of how the triggered DWFS snapshots are generated, see Section [17](#page-46-1).

Both TM_LF_DWF and TM_LF_DSWF can be generated at the same time and can be configured, enabled or disabled independently.

16.4 Spectral data products (TM_LF_SM, TM_LF_BP0-2)

The LF board FPGA provides to the DPU averages spectral matrices. These are further processed in the software. Several types of post processing will be applied to SM packets to produce different spectral products:

16.4.1 TM_LF_SM data product:

This product contains the full spectral matrices, further averaged in software and converted to a more efficient format.

- Can be produced in parallel with BP0, BP1 or BP2.
- Selection of components: Typically, not all 8x8 components are valid/useful and the software will reduce the matrix to a subset of e.g. 6x6 or 3x3. These components are selected by the *SM_COMPS* bitmask.
- Reduction of the 64 bits integers to a more compressed format. The diagonal elements of the matrix are compressed to a 16-bit float value and each pair of the non-diagonal elements (forming a complex number) are normalized to an integer complex number encoded in a 16-bit integer (2 x 8 bits).

16.4.2 TM_LF_BP0 data product:

This product contains integrated power spectra calculated from the SM, summed over several components

- BP0 can be produced in parallel with TM_LF_SM but not with BP1/BP2. BPx products are mutually exclusive.
- The data product contains two averaged auto spectra (E and B):
- These are calculated as summed auto-spectra (diagonal elements of the SM). Summation is performed according to the bitmask specified in *BP_MASK_EB*.
- Spectra can be reduced by averaging in time and frequency according to *BP_AVG_TIME* and *BP_AVG_FREQ_LOG2*.

16.4.3 TM_LF_BP1 data product:

Calculation reduced version of spectral parameters (reduced spectral matrix + Poynting vector). Also includes in the header the magnetic field vector provided to RPWI by JMAG. The calculation is performed by

- Averaging of transformed spectral matrices in time and frequency (set via *BP_AVG_TIME* and *BP_AVG_FREQ_LOG2).*
	- Processing of averaged spectral matrices to obtain, for each frequency step:
		- three components of E-field spectrum encoded as 3×8 bits mantissa + a common 6-bit exponent
		- three components of B-field spectrum encoded as 3×8 bits mantissa + a common 6-bit exponent
		- Off-diagonal elements
		- Poynting vector : 3x8=24 bit,
		- \rightarrow Total set: 128 bit (16 byte) per frequency and time interval.

16.4.4 TM_LF_BP2 data product:

This product contains the wave polarization parameters (power, ellipticity, planarity and Poynting vector direction) in a highly compressed form. Again, this product can be generated in parallel with TM_LF_SM, but not with other BPx products. The calculation involves transforming the spectral matrices to field aligned coordinates using JMAG data, averaging, applying on-board calibration and afterwards calculating the spectral parameters.

Details of the spectral calculation are given in an appendix (Section [18](#page-50-0)).

16.4.5 Configuration of spectral bins and include masks

As described in Section [13](#page-22-1), the LF FPGA calculates the spectrum with 2048 FFT, yielding 1024 "source" frequency bins, and afterwards accumulates the spectral matrices both in time and frequency. All spectral data use a common "frequency" axis (specifying the output frequency bins after frequency reduction) and include mask (which allows to include/exclude individual bins of the source 1024 point spectrum – masked out bins, those set to 0 in the mask, are excluded from averaging/summing).

In the PC configuration, the index *SM_BINEDGES_TABLE_INDEX* gives and index to the active SB_TABLE (a table containing multiple sets of frequency bins – see [14.2](#page-28-0)) and *SM_INCLUDE_MASK_INDEX* is an index to the active EFM_TABLE, where the include masks are specified as 1024bit bitmasks.

The number of output frequency bins for the TM_LF_SM data product must be configured in *SM_NUM_FREQ_BINS*.

For BP0/BP1/BP2 products, the same frequency axis is used, but may be further reduced in frequency by summing adjacent frequency bins together. This frequency reduction is configured by

BP_AVG_FREQ_LOG2. The software sums together 2^*BP_AVG_FREQ_LOG2* adjacent bins, so for BP_AVG_FREQ_LOG2 = 0, the BPx frequency axis is the same as SM axis, for BP_AVG_FREQ_LOG2 = 1, the number of frequency bins in BPx products is reduced to one half, comparing to SM.

16.4.6 Handing of reaction wheel data

LF software receives reaction wheel information from the OBC and dynamically updates the include mask in the FPGA to mask out the RW frequencies. There are 4 reaction wheels and the software allows to mask out the frequency f and its 8th harmonic 8^{*}f. The frequencies to mask are configured by the **RW_MASKING**

parameter in PCE configuration (bit 0 = fundamental of wheel 1, bit $1 - 8$ th harmonic of wheel 1, bit 2 fundamental of wheel 2, ….).

Once the LF SW receives a new value for the RW frequencies, it calculates the frequencies to mask and applies a new mask on the start of the next LF_SM data product. This new mask is constructed from the static mask specified by *SM_INCLUDE_MASK_INDEX* with additional frequencies masked due to RW. The fact that a new mask has been applied is indicated in the TM_LF_SM/TM_LF_BPx packet headers by bit 15 in NUM_AVG_SPEC. The packet headers also reflect the masked frequencies in the 8-byte field RW_FREQS.

16.5 Interaction with RPWI software sequencer

The LF hardware does not require real time control from the sequencer and runs autonomously. In usual operation modes, the sequencer / software initializes the LF to a given configuration and LF operates in this mode autonomously, independently of the sequencer cycle. However, the LF software handles the following sequencer events on software level:

RPWI_EV_INTERFERENCE_START and RPWI_EV_INTERFERENCE_STOP: These two events signal to the LF software that E-field data is most likely affected by LP or MIME sweeps and do not contain a valid natural signal. LF software reacts in the following manner:

- 1) It pauses averaging of spectral matrices and BPx data products. During this interval, no new datapoints are included in the spectral averaging.
- 2) LF internal triggering in (in LF_ALGO_DUST_WAVE and LF_ ALGO_BP2 modes) is paused, so the SW will not trigger inside this interval.

RPWI_EV_TRIGGER: This event is used only in triggering mode based on LP detection

(LF_ALGO_EXTERNAL). In this mode, LF software triggers a snapshot on reception of the event.

RPWI_EV_CONCLUDE_TRIGGER: Used always in LF_ALGO_EXTERNAL mode and in in

LF_ALGO_DUST_WAVE and LF_ALGO_BP2 modes when **TRIG_ALGO_CONCLUDE** bit is set in configuration. In this configuration, the SW dumps the content of the temporary buffer only when this event is received, not autonomously.

RPWI_EV_INIT: LF software performs static initialization

RPWI_EV_CONFIG: LF applies a given configuration and starts LF science operation

RPWI_EV_STOP and RPWI_EV_ABORT_R: When either of those events is received, RPWI SW resets the LF board.

17 Triggering, dust detection and statistics (TM_LF_STAT)

The LF software allows multiple modes of triggering of its triggered snapshots (TM_LF_TSWF and TM_LF_DWFS). The global modes of LF trigger operation is specified by TRIG_ALGO variable in the PCE structure. The ALGO_CODE can be set to one of the following:

- LF_ALGO_EXTERNAL: LF uses external trigger from the LP board, controlled by the sequencer. This is the same operation as was implemented in Release 1.
- LF_ALGO_DUST_WAVE: LF internal triggering, when LF triggers autonomously based on the 49 ksps data it collects. It attempts to detect waves and dust impacts in the data via a set of criteria. In this regime (and only in this regime), the TM_LF_STAT product is generated.
- LF_ALGO_BP2: Triggers the waveform snapshots based on spectral data, calculated by LF.

17.1 Basic triggering using LP detection (LF_ ALGO_EXTERNAL)

In this mode, LF triggers using the EV_TRIGGER and EV_CONCLUDE_TRIGGER events received from the sequencer, based on LP data. This mode only works with the sequencer, if run without a sequencer, no triggered snapshots are transmitted.

The basic operation is as follows:

- Whenever LF receives the EV_TRIGGER from LP, it commands the FPGA to capture TSFW snapshot (if TSWF enabled) and stores the snapshot in its internal buffer.

- A DWFS snapshot is captured as well (if DWFS product is enabled and the *CFG_EXTRAS_DWFS_NO_SEQ* bit is not set), centered on the time of EV_TRIGGER and stored in a separate temporary buffer.

- If a new trigger EV_TRIGGER arrives, the buffers are overwritten.

- When EV_CONCLUDE_TRIGGER is received the content of the temporary buffers is transmitted in the form of TM_LF_TSWF and TM_LF_DWFS products and the buffers are emptied. If the buffers are empty, the snapshots are taken at the time of EV_CONCLUDE_TRIGGER and transmitted immediately.

17.2 Autonomous triggering with dust detection (LF_ ALGO_DUST_WAVE)

The triggering is performed by analyzing waveform snapshots taken periodically by LF.

- The source snapshot length and cadence is configured in PCE structure via *wfs_length* and *wfs_period* parameters.
- Triggering is performed on one component which is selected by *trig_channel* variable in PCE structure.

Below is described the algorithm (Note: the *bold-italic* variables in this section correspond to configuration parameters from the PCE structure):

Inputs: snapshot E(t), N = *wfs_length**128 samples, triggering done based on a selected component

- **1) For each snapshot we compute:**
- $MAX = max(abs(E))$
- $MED = median(abs(E))$
- $ZX =$ number of zero crossings of $(E **trig** ZX)$ offset offset). Proportional to frequency for narrowband waves.
- RMS $E =$ RMS value of E
- RMS B = RMS value of B (summed over alt channel bitmask specified by bits in **trig_alt_channel_mask**)

2) Snapshot is classified as either wave or dust

if ((MAX/MED > *trig_thr_ratio_dust*) & (ZX < Thresh3) & (RMS_B < Param_A)) **=> the snapshot is dust** else if ((MAX/MED < Thresh2) & (ZX > Thresh4) & (RMS_B > Param_B)) **=> the snapshot is a wave**

3) Quality factor calculation:

A quality factor **Q** is calculated from the snapshot. A method is chosen by the *trig_quality* variable and can be one of:

- RMS_AMP largest rms amplitude signal stored
- PEAK_AMP largest peak amplitude signal stored
- RATIO largest MAX/MED ratio
- ZEROX -- number of zero crossings
- DUST_PEAK peak amplitude with preference to dust (+ 32k if dust)
- DUST_RATIO RATIO with preference to dust (+ 32k if dust)
- WAVE_PEAK peak amplitude with preference to waves (+ 32k if waves)
- WAVE_RMS -
- RMS_ALR_CH RMS value from alternate channels (summed).

4) Triggered snapshot selection

The algorithm keeps the highest quality snapshot in an internal buffer (unless immediate triggering is configured), until a higher quality snapshot is found. If the current snapshot has a higher **Q** than the stored one, the stored one is replaced by the new one. The triggered snapshot is transmitted as TM_LF_TSWF and the internal buffer is cleared using one of the approaches defined in section [17.4.](#page-49-1)

*DFWS triggering:*In this triggering mode the DWFS snapshot can either be triggered at the same time as the TSWF (the DWFS trigger following the TSWF trigger) or the DWFS triggering can follow the basic mechanism described in [17.1](#page-47-0), when DWFS is triggered by the EV_TRIGGER event from sequencer.

5) Statistic (TM_LF_STAT) calculation.

The results of the processing of all the waveform snapshots processed by the algorithm are used to collect statistics. The results are transmitted in the form of STAT blocks, with each block containing information collected over *TRIG_NUM_SNAP_STAT* processed waveform snapshots. This includes the number of positive and negative dust spikes identified, number of waves identified, maximum and average signal amplitude etc. (see the TM_LF_STAT packet description). To reduce packet overhead, multiple STAT blocks can be combined in a single TM_LF_STAT telemetry packet. The number of blocks in a packet is set by *STAT_BLOCKS_PER_PACK.*

17.3 Triggering based on BP2

In this regime, the trigger for the TSWF triggered snapshot is based on a quality factor calculated from the BP2 spectral product. This importantly requires that WFS and BP2 products are configured synchronously (WFS snapshot starts at the same time as BP2 spectral data collection). So, the WFS_PERIOD must match BP_AVG_TIME and NAS setting.

The following algorithm is then used to calculate the quality factor which is then used for triggering.

```
quality = 0;
for (idx = trig_bp2_index_low; idx < trig_bp2_index_high; idx++) 
{
   bool BtraceOK = (Btrace >= trig_bp2_b_low) && (Btrace < trig_bp2_b_high)
   bool EtraceOK = (Etrace >= trig_bp2_e_low) && (Etrace < trig_bp2_e_high)
   bool ElipOK = …. and analogously for planarity, theta, phi, Sz.
   if (BtraceOK && EtraceOK && ElipOK && ......) {
            quality += Btrace 
    }
}
if (quality > trig_bp2_q_threshold) {
        trigger the snapshot!
}
```
*DFWS triggering:*In this triggering mode the DWFS snapshot can either be triggered at the same time as the TSWF (the DWFS trigger following the TSWF trigger) or the DWFS triggering can follow the basic mechanism described in [17.1](#page-47-0), when DWFS is triggered by the EV_TRIGGER event from sequencer.

17.4 Common additional triggering settings (handling of long snapshots, immediate trigger, dump cycling)

For all the triggering algorithms (LF_ ALGO_DUST_WAVE and LF_ ALGO_BP2), a common mechanism is used to determine when is the triggered snapshot sent out (and the internal buffer cleared in applicable). This general mechanism is described here together with some specifics.

In normal operation, the currently "best" triggered snapshot is kept in memory buffer and is sent to the spacecraft after the end of the current cycle (ether driven by sequencer or by internal LF cycle – see below). This is however only possible for snapshots up to than 96 ksamples per component, due to memory

limitation of RPWI. To allow LF to transmit longer snapshots, up to 256 ksamples a specific mechanism must be enabled as follows (This can be used with all LF_ALGO* algorithms):

- CFG_EXTRAS_WFS_EX_LONG bit has to be set in the *EXTRA_SETTINGS* to allow for the snapshot length to exceed 32 ksamples (per channel). When this is set, the RSWF and TSWF snapshots are splits in blocks of 32 ksamples. This is also requires the snapshot length specified in WFS_LENGTH to be a multiple of 32 ksamples.
- If snapshots longer than 96 ksamples are required, the TRIG_ALGO_IMMEDIATE bit must be set as well. In this case, the triggered snapshots are not kept in the internal temporary buffer, but are immediately transmitted to the DPU. To avoid overloading telemetry in case of frequent triggers, the *TRIG_ALGO_LIMIT* bit should be set, limiting the number of triggered snapshots to be sent to one per *TRIG_DUMP_CYCLE*.

For internal triggering algorithms (LF_ ALGO_DUST_WAVE and LF_ ALGO_BP2), additional trigger behavior is determined by the setting of *TRIG_ALGO_CONCLUDE* bits in configuration:

- if TRIG_ALGO_CONCLUDE is set, the snapshot is dumped when EV_CONLCUDE_TRIGGER is received from the sequencer.
- if **TRIG_ALGO_CONCLUDE** is not set, the LF dumps the data autonomously after it processes *TRIG_DUMP_CYCLE* snapshots.

18 Appendix 1: BP2 spectral parameter calculation description

This was taken from an internal specification, needs clarification.

18.1 **Configuration parameters:**

- **nBins** = number of frequency bins in spectral matrix.
- **indicesED** = indices of E antennas to use in Poynting calculation (there can be up to 5 antennas)
- **bp_avg_time** and **bp_avg_freq_log2** are used the same way as for BP0 and BP1
- **indicesTrace** = indices of E antennas to use in calculation of the trace of Electric field (1,2 or 3 antennas can be selected)

Note: indices start at 1 in this section.

18.2 **Calibration tables/matrices:**

These are defined in LF_Tables.h/.c

1) **TM_scm_sc (lfTransMatB in code):** 3x3 real matrix, transforming from SCM coordinates to SC coordinates. For example

B_sc = **TM_scm_sc** * **B_scm** (B_scm is SCM magnetic field in original sensor coordinates)

2) **Ant_dir_sc (lfTransMatEAnt in code):** 3 x 5 real matrix, columns are unit directions of antennas corresponding to E1,E2,…, E5 in SC coordinates

3) **CalMat** (8 x 64 complex – **lfSmCalMatrices in code**), read from lfSmCalMatrices. Frequency dependent calibration of 8 components for a given mode at 64 frequency bins. Used to generate calibration matrices to convert the integer M_fpga matrix to physical units.

4) **TM_ant_sc (lfTransMatE in code):** 3x3 matrix corresponding to a chosen indicesTrace. A matrix to convert a 3 component E-field vector from antenna coordinates to SC coordinates. Will include antenna lengths. The matrix is computed on ground and uploaded to SW. Software should have a set of 4 such matrices, selectable by a TC

5) **TM_mag_sc (lfTransMatJmag in code):** Transformation matrix from JMAG sensor coordiates to SC coordiates.

18.3 **Definitions:**

M_fpga (matrix 8 x 8 x nbins, 64bit integer) – matrix from FPGA, averaged in frequency in SW if configured. **EB(i,j,f)** [3 x 5 x nbins, complex float] – submatrix of ExB elemens of M_fpga, calibrated

BB(i,j,f) [3 x 3 x nBins, complex float] – submatrix of BxB elemens of M_fpga, calibrated

TM_sc_mfa [3 x 3 x nBins, real float] – transformation matrix from SC to field aligned (MFA) coordinates, where

- axis mfa3 is along B_jmag

- axis mfa1 is orthogonal to 3, ad lies in a plane containing X axis of SC system.

- axis mfa2 is orthogonal to mfa1 and mfa3

TM_sc_mfa is calculated as (where x is Vector cross product):

 $T(3,:) = B_jmag / norm(B_jmag);$ $T(2;) = B_j$ mag x $[1,0,0] / norm(B_j)$ mag x $[1,0,0]$ $T(1,:) = T(2,:) \times T(3,:)$

18.4 **Initialization steps (to be done on EV_CONFIG, before first matrix calculation)**

Step 1: **CalMatInt** = CalMat, interpolated to actual frequency bins (nbins). Can be done "virtually" in a function to save memory.

Step 2: Precalculate CM_BE: 3 x 5 x nBins complex calibration matrices for BE submatrix

CM_BE($i=1..2$, $j=1..5,f$) = CalMatInt (i,f) *conj(CalMatInt($j+3,f$)), where $f = 1..n$ bins;

Step 3: Precalculate CM_BB: 3 x 3 x nBins calibration matrices for BB submatrix and EE submatrix

// Calibration matrix for BB submatrix

CM_BB($i=1..3$, $j=1..3, f$) = CalMatInt (i,f) *conj(CalMatInt (j,f)), where $f = 0..$ nbins-1;

 \textit{l} CM_EE: 3x3 calibration matrix for trace. Only needed if (length(indicesTrace) == 3) **CM_EE**(i=1..3.j=1..3,f) = CalMatInt(3+indicesTrace(i),f)*conj(CalMatInt(3+indicesTrace(j),f))

// Calibraiton coefficients for diagonal of E

CM_Ediag($i=1..5,f$) = CalMatInt($i+3,f$)*conj(CalMatInt($i+3,f$))

Step 4: Initialize averaged matrix BB_avg(i,j,f), trace of E components E_trace(f) and Poynting component Sz(f) to zero.

18.5 **Routine processing, for every SM matrix received:**

Step 1: From FPGA, SW receives an a matrix **M_fpga** (8 x 8 x nbins). Apply averaging in frequency, if applicable to get "nBins" frequency bins.

Step 2: Generate calibrated sub-matrices EB and BB

BB(i,j,f) = M_fpga(i,j,f) * CM_BB(i,j,f), for all i, j, f - element-wise multiplication $BE(i,j,f) = M$ fpga(i,3+j,f)*CM_BE(i,j,f), for all i, j, f - element-wise multiplication

And diagonal of calibrated E-field matrix:

E diag(i,f) = M fpga(3+i,3+i,f)*CM Ediag(i,f);

Step 3: Software waits until JMAG B field, spanning the time when M_fpga was averaged, is available.

B_jmag_mag = JMAG vector averaged over this interval (in MAG instrument coordinates).

Step 4: Using B_jmag, calculate **TM_sc_mfa**

// Convert JMAG B from JMAG coordinates to SC coordinates

B_jmag_sc = TM_mag_sc* B_jmag_mag;

// Create transformation matrix

 $T(3,:) = B$ jmag_sc / norm(B_jmag_sc); $T(2;)=B_j$ mag_sc x $[1,0,0]$ / norm (B_j) mag_sc x $[1,0,0]$) $T(1,:) = T(2,:) \times T(3,:)$ **TM_sc_mfa =** T**;**

Step 5: Calculate **TM_scm_mfa** = TM_sc_mfa*TM_scm_sc;

Step 6: Transform B components and antenna directions to MFA

for all **f** do:

BB_mfa = TM_scm_mfa*BB*conj(TM_scm_mfa)

BE_mfa = TM_scm_mfa*BE

Ant_dir_mfa = TM_sc_mfa*Ant_dir_sc; // Ant_dir_* are unit vectors

Step 7: Calculate projections of Poynting vectors to JMAG B-field direction (Z-axis in MFA)

numavg = number of averaged FFTs in FPGA (constant);

```
for all frequencies f do:
  Sz(f) = 0;
   for i in <indicesED>:
     for i=1,2// Calculate variance (sigma^2) of each element of BE_mfa
       // Use a formula from Priestley et al. (p. 702)
        var(BE_mfa(j,i)) = BB_mfa(j,j)*E_diag(i) + real(BE_mfa(j,i))^2 - imag(BE_mfa(j,i))^2;
      end
      // calculate a projection of the Poynitng vector to MFA Z-axis
     // Ant_dir_mfa(i,j) is the i-th component of the unit direction vector of j-th antenna in MFA
      Sz_proj = real(BE_mfa(2,i)) * Ant_dir_mfa(1,i) - real(BE_mfa(1,i)) *Ant_dir_mfa(2,i) 
      // calculate a normalization factor
     normf = sqrt((var(BE_mfa(2,i))*Ant_dir_mfa(1,i)^2 + var(BE_mfa(1,i)) * Ant_dir_mfa(2,i)^2) /
numavg);
      // accumulate the projections for all antennas
     Sz(f) += Sz proj / normf;
    end
```
end

- 55/61 -

Step 8: Add to averaged values

for all f,

```
BB_avg(:,:,f) += BB_mfa(:,:,f);
```
Sz $avg(f) += Sz(f);$

// EE_avg64 (I64 3x3 matrix): akumuluje se nekalibrovana 3x3 sub-matice EE podle indicesTrace

EE_avg64(:,:,f) += M_fpga(3+indicesTrace,3+indicesTrace,f);

endfor

18.6 **Final processing after averaging of all matrices is completed:**

Step 1: Calculate E_trace: Trace of E components averaged. If exactly 3 E components are selected, transform to orthogonal components before computiong trace.

for all f do:

```
EE_avg(i,j) = EE_avg64(i,j,f) * CM_EE(i,j,f), for all i, j
 len = length(indicesTrace);
if (len < 3)
   E_trace(f) = sum(EE_avg(i,i,f)), for i=0:len-1
 else // only if (length(indicesTrace) == 3)
   EE_trans = TM_ant_sc*EE_avg*conj(TM_ant_sc);
   E_trace(f) = sum(EE_trans(i,i,f)), for i=0:len-1
 end
```
Step 2: Generate a packet with the following 4-byte structure for every frequency bin:

// Total number of averaged FFT spectra (FPGA & SW averaging combined) tot_num_avg = numavg*num_averaged_matrices;

B_trace = trace(BB_avg)/tot_num_avg, converted to 8 bits via Util_LogEncode32to8

E_trace = E_trace/tot_num_avg, converted to 8 bits via Util_LogEncode32to8

Run SMX_Prassadco(BB_avg) to get:

```
float k_vector[3], ellipticity, planarity (ellipticity = -1 to 1, planarity 0..1, |k_vector| = 1)
```
Reduce to the following structure:

theta (4 bits 0-90deg, resolution 5.625deg): Angle between of k_vector and B0 magnetic field.

Theta [radians] = $arccos(abs(kz))$; // assuming k_vector = [kx, ky, kz] a |k_vector| = 1

Phi (4 bits, -180-180deg, 22.5deg resolution): Azimuthal angle of k_vector "around" B0.

Phi [radians] = atan2(ky, kx); // check if atan2 works OK if ky == 0 or kx == 0

ellipticity (3bits, signed -1..1), independent of B0, step 0.25

Planarity (3bit, unsigned, between 0-1), step 0.125, independent of B0

Sz_avg (2bit): Sz_avg/num_averaged_matrices reduce to 2 bits by comparing absolute value to a defined threshold (set in config) and keeping the sign. If below threshold \Rightarrow 0 or -1, if above threshold \Rightarrow -2 or 1.

Reset BB_avg, Sz_avg and EE_avg64 to zeros and start new averaging

19 Appendix 2: Defaults configurations in SW2.0

19.1 Built-in LF configurations

The table below lists the default LF configurations hard-coded in SW2.0.

JUICE RPWI LF digital interface

 $\begin{array}{c} \hline \end{array}$

Ref: JUI-IAP-RPWI-LF-DIF Revision: 154 Date: **04/10/2023**

- 58/61 -

19.2 Standard E-field mux configs used in the above LF configurations

Table 19-2: Configuration of multiplexers used in standard configurations. Highlighted components are the ones which are usually transmitted (others are filtered by software)

 $\overline{}$

19.3 Use of LF configurations in RPWI operational sequences

This table lists the proposed default assignment of LF configurations to the LF sequences.

